

B1:

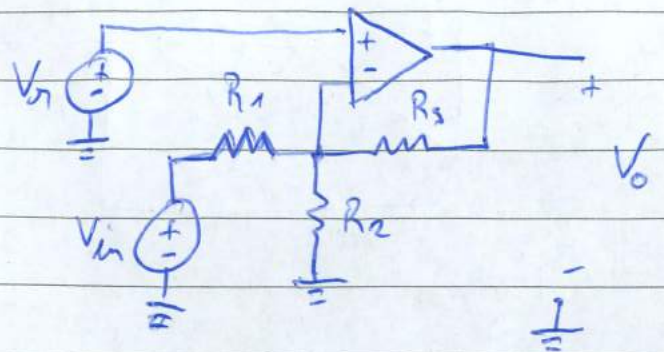
$$V_+ = V_- = V_-$$

KCL on V_- gives:

$$\frac{V_{in} - V_-}{R_1} - \frac{V_-}{R_2} - \frac{V_- - V_o}{R_3} = 0 \quad (=)$$

$$\frac{V_o}{R_3} = + \left(\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right) V_- - \frac{1}{R_1} V_{in} \quad (=)$$

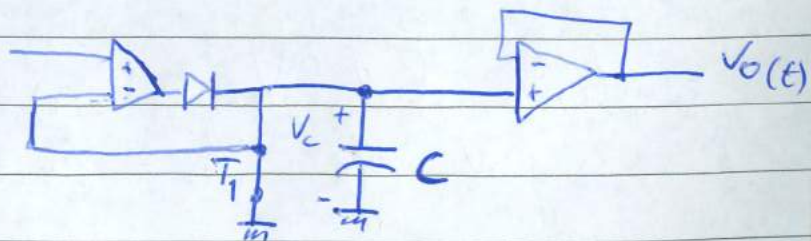
$$V_o = \left(1 + \frac{R_1 R_2 R_3}{R_1 + R_2} \right) V_- - \frac{R_3}{R_1} V_{in}$$



B2:

The Key is to analyze the circuit for different states of the semiconductors D_1 and T_1 .

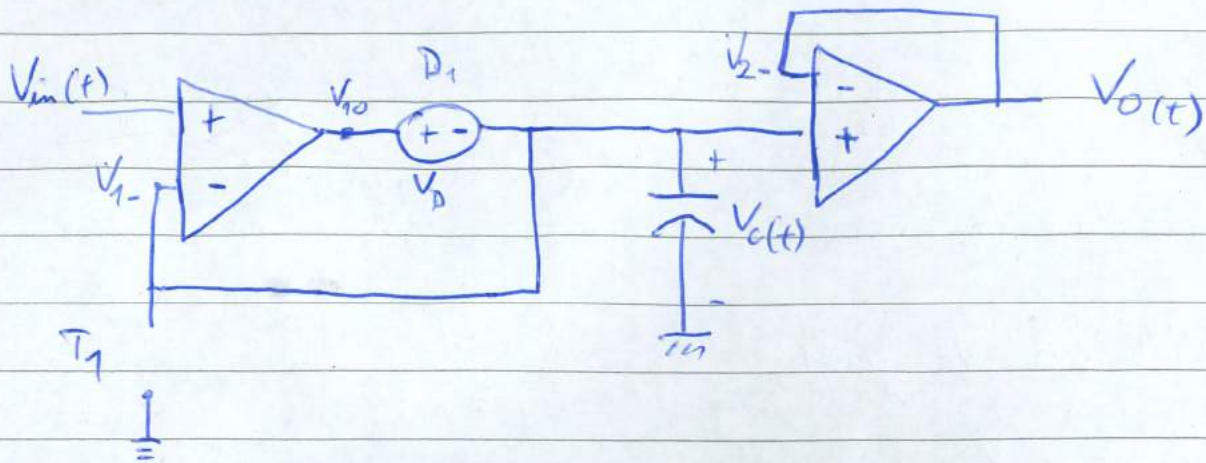
i) T_1 is ON:



The capacitor is short-circuited to ground, so $V_c = 0$ and $V_o(t) = V_c = 0$.

This happens independently of $V_{in}(t)$ and the state of D_1 . And this state is valid when $V_{in}(t) > 2V$

ii) T_1 is OFF & D_1 is ON:



Due to the negative feedback of the OpAmps:
 $V_- = V_+ \Rightarrow V_{C(t)} = V_1^- = V_{in}(t)$

$$V_O(t) = V_{2-} = V_{C(t)}$$

$$V_O(t) = V_{in}(t)$$

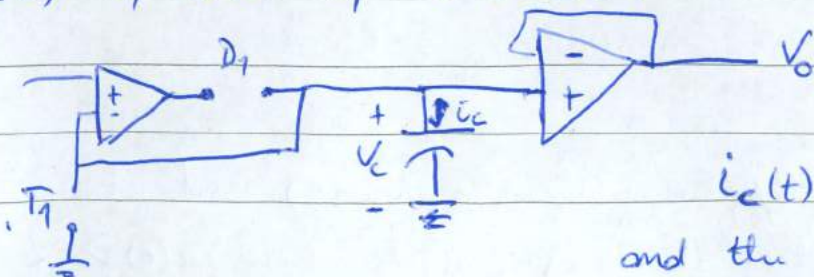
For this to be valid, we must have

$$\underbrace{V_{10}(t) < V_{GS,th} = 2V}_{(T_1 \text{ OFF})} \quad \text{and} \quad \underbrace{V_D = V_{10} - V_{C(t)} > 0.7V}_{(D_1 \text{ ON})}$$

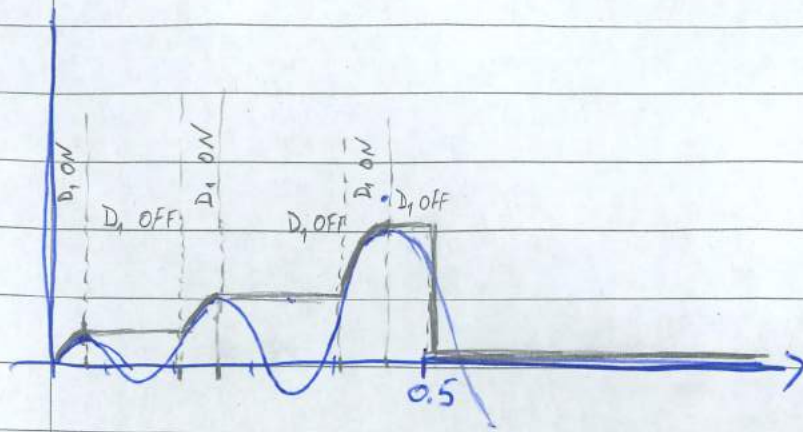
Since $V_{10} = A_{10}(V_{1+} - V_{1-})$ and $A_{10} \approx \infty$,
 we get that D_1 is ON if $V_{1+} - V_{1-} > \frac{V_{C(t)} + 0.7}{A_{10}} \approx 0$

Recalling that $V_{1-} = V_{C(t)}$, we conclude that
 D_1 is ON if $V_{1+} = \underline{V_{in}(t)} > V_{C(t)}$

iii) T_1 and D_1 are OFF:



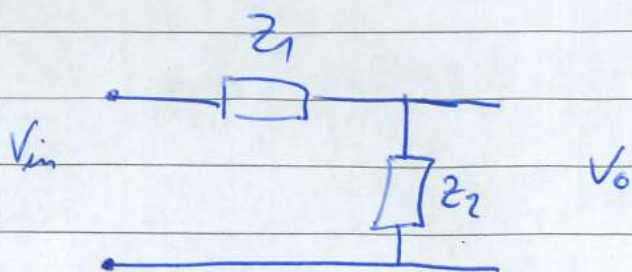
$i_C(t) = 0$, so $\frac{d}{dt} V_{C(t)} = 0$
 and the voltage will be kept constant.



T_1 is OFF T_1 is ON

33

We want a BP filter. The filter will be of the form:



where Z_1 and Z_2 will be determined by arranging one R, one L, and one C.

$$V_o = \frac{Z_2}{Z_1 + Z_2} V_{in} = H(j\omega) V_{in}$$

$Z_{eq} = Z_1 + Z_2$ is the total impedance, and it will remain the same regardless of how we arrange R, L and C.

$$Z_{eq} = R + \frac{1}{j\omega C} + j\omega L$$

$$H(j\omega) = \frac{Z_2}{R + \frac{1}{j\omega C} + j\omega L}$$

BP filter means: $H(0) = 0$, $H(\infty) = \infty$

$$i) H(0) = \frac{Z_2(0)}{R + \frac{1}{j\omega C} + j\omega L} \approx \frac{Z_2(0)}{\frac{1}{j\omega C}} = j\omega C Z_2 = 0$$

This means Z_2 cannot have C in it!

$$ii) H(\infty) = \frac{Z_2}{R + \frac{1}{j\omega C} + j\omega L} \approx \frac{Z_2}{j\omega L} = 0. \text{ so } Z_2 \text{ cannot have L in it!}$$

$$\therefore Z_2 = R$$

